

## High Precision, Rail-to-Rail Input and Output Operational Amplifier

### Overview

SL8551 (single), SL8552 (dual) and SL8554 (quad) are high-precision, rail-to-rail input and output operational amplifiers. The chip uses offset correction technology, so it has very low offset voltage (typical value  $2\mu\text{V}$ ), and the offset voltage hardly changes with temperature and time.

The chip supports single power supply and dual power supply. When powered by a single power supply, its power supply voltage range is  $+2.3\text{V}$  to  $+5.5\text{V}$ ; when powered by a dual power supply, its power supply voltage range is  $\pm 1.15\text{V}$  to  $\pm 2.75\text{V}$ .

The package types of SL8551 are mainly SC70-5, SOT23-5, MSOP-8 and SOP-8. The package types of SL8552 are mainly MSOP-8 and SOP-8. The package type of SL8554 is mainly SOP-14. And the operating temperature range for all package types is  $-40^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ .

### Features

- Low offset voltage:  $2\mu\text{V}$  (typical value)
- Zero drift:  $0.03\mu\text{V}/^{\circ}\text{C}$
- Low noise:  $30\text{nV}/\sqrt{\text{Hz}}$   
0.1Hz to 10Hz : noise  $0.55\mu\text{VPP}$
- High DC accuracy:  
Open loop gain: 135dB  
Power Supply Rejection Ratio (PSRR): 110dB  
Common Mode Rejection Ratio (CMRR): 110dB
- Gain bandwidth product: 2MHz
- Quiescent current:  $220\mu\text{A}$  (typical value)
- Supply voltage range:  $\pm 1.15\text{V}$  to  $\pm 2.75\text{V}$
- Input and output rail to rail

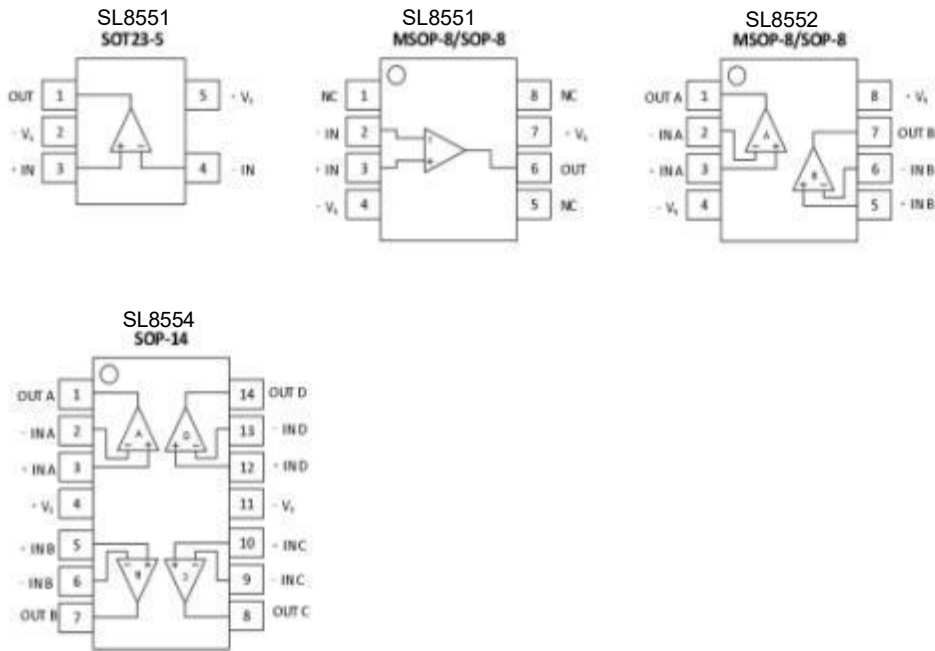
### Application

- Bridge amplifier
- Strain gauge
- Sensor applications
- Temperature measurement
- Electronic scale
- Medical devices

## Application

- Resistance temperature sensor
- Handheld test equipment

## Pin distribution



Pin distribution

## Pin description

Pin number	Symbol	Description
1	-IN	The inverting input of the operational amplifier has an input voltage range from (VS-) to (VS+).
2	+IN	The non-inverting input of the operational amplifier has the same input voltage range as the inverting input.
3	+Vs	Positive power supply terminal, whose voltage range is 2.3V to 5.5V ( $\pm 1.15V$ to $\pm 2.75V$ ).
4	-Vs	Negative power supply terminal, connected to ground when single power supply is used.
5	OUT	Output of the operational amplifier.
6	N/C	No connection.

## Ordering information

Model	Packing	Boxing quantity
SL8551XC5	SC70-5	Reel 3000 PCS
SL8551XT5	SOT23-5	Reel 3000 PCS
SL8551XS8	SOP-8	Reel 4000 PCS
SL8551XV8	MSOP-8	Reel 3000 PCS
SL8552XS8	SOP-8	Reel 4000 PCS
SL8552XV8	MSOP-8	Reel 3000 PCS
SL8554XS14	SOP-14	Reel 2500 PCS

## Absolute maximum ratings (ambient temperature 25°C)

Symbol	Parameter	Rating	Unit
Supply voltage		+3 , +6 (Single power)	V
Single input	Voltage	$V_s - 0.5$ to $V_{s+} + 0.5$	V
	Differential voltage	$\pm 5$	V
Temperature range	Operating temperature <sup>(2)</sup>	-55 to 150	°C
	Storage temperature, $T_{STG}$	-65 to +150	
	Junction temperature, $T_J$	150	
Electrostatic discharge (ESD)	Human model (HBM)	8	kV

Note:

1. Exceeding the absolute maximum ratings may cause permanent damage to the device. The above listed parameters are only some of the key parameters, and do not mean that other parameters not listed can exceed the normal range of use. Long-term operation at the absolute maximum ratings may affect the reliability of the device;
2. The device cannot exceed the maximum junction temperature at any time;

## Electrical parameters

$V_S=+5V$  ,  $T_A=+25^{\circ}C$  ,  $V_{CM}=V_S/2$  ,  $V_O=V_S/2$  ,  $R_L= 10k\Omega$  , connect to  $V_S/2$  , unless otherwise stated.

Symbol	Parameter	Condition	Min	Typ	Max	Unit
Input characteristics						
$V_{OS}$	Input offset voltage			2	15	$\mu V$
$V_{OSTC}$		$T_A=-40^{\circ}C$ to $+125^{\circ}C$		0.02		$\mu V/^{\circ}C$
$I_B$	Input bias current	$V_{CM}=V_S/2$		$\pm 100$		PA
$I_{OS}$	Input offset current			$\pm 100$		PA
$V_{CM}$	Common mode input voltage range	$T_A=-40^{\circ}C$ to $+125^{\circ}C$	$V_{S-}$		$V_{S+}$	V
CMRR	Common mode rejection ratio	$V_{S-}<V_{CM}<V_{S+}$	90	110		dB
		$T_A=-40^{\circ}C$ to $+125^{\circ}C$	85			dB
$A_{VOL}$	Open loop voltage gain	$V_{S-}+0.3V<V_O<V_{S+}-0.3V$	105	135		dB
		$T_A=-40^{\circ}C$ to $+125^{\circ}C$	100			dB
Output Characteristics						
$V_{OH}$		$R_L= 10K\Omega$	$(V_{S+})-12$	$(V_{S+})-4$		mV
		$T_A=-40^{\circ}C$ to $+125^{\circ}C$	$(V_{S+})-8$			mV
$V_{OL}$		$R_L= 10K\Omega$		$(V_{S-})+4$	$(V_{S-})+12$	mV
		$T_A=-40^{\circ}C$ to $+125^{\circ}C$			$(V_{S-})+18$	mV
$I_{SC}$	Short circuit output current	Source current	55	65		mA
		$T_A=-40^{\circ}C$ to $+125^{\circ}C$	50			mA
		Sink Current	48	55		mA
		$T_A=-40^{\circ}C$ to $+125^{\circ}C$	45			mA
Power supply characteristics						
PSRR	Power Supply Rejection Ratio	$V_S=2.3V$ to $5.5V$	90	110		dB
		$T_A=-40^{\circ}C$ to $+125^{\circ}C$	80			dB
$I_Q$	Quiescent current			220	290	$\mu A$
		$T_A=-40^{\circ}C$ to $+125^{\circ}C$			380	$\mu A$
Noise characteristics						
$e_n$	Input voltage noise	$f=0.1Hz$ to $10Hz$		550		nVpp
	Input voltage noise	$f=1KHz$		30		$nV/\sqrt{Hz}$
Dynamic characteristics						
GBW	Gain bandwidth product			2		MHz
SR	Slew rate	$G= \pm 1$		0.8		$V/\mu s$
$t_{OR}$	Overload recovery time	$V_{IN}\times G=V_S$		50		$\mu s$

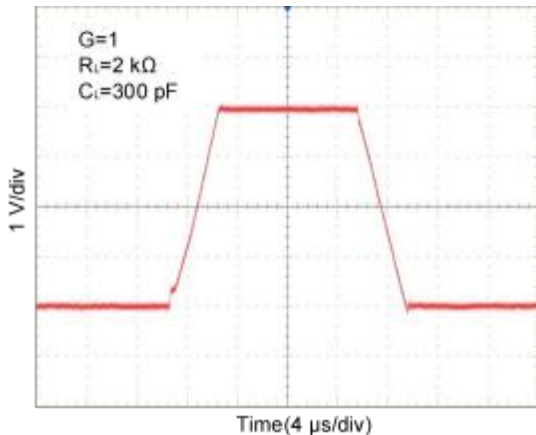
## Electrical parameters

$V_S = \pm 2.7V$ ,  $T_A = +25^\circ C$ ,  $V_{CM} = V_S/2$ ,  $V_O = V_S/2$ ,  $R_L = 10k\Omega$ , connect to  $V_S/2$ , unless otherwise stated.

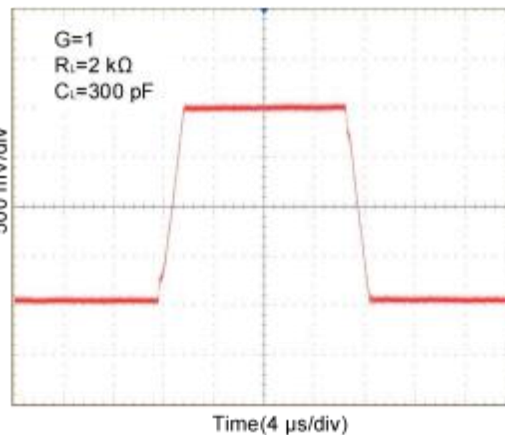
Symbol	Parameter	Condition	Min	Typ	Max	Unit
Input characteristics						
$V_{OS}$	Input offset voltage			2	15	$\mu V$
$V_{OS}TC$		$T_A = -40^\circ C$ to $+125^\circ C$		0.02		$\mu V/^\circ C$
$I_B$		$V_{CM} = V_S/2$		$\pm 100$		PA
$I_{OS}$	Input offset current			$\pm 100$		PA
$V_{CM}$	Common mode input voltage range	$T_A = -40^\circ C$ to $+125^\circ C$	$V_{S-}$		$(V_{S+})$	V
CMRR	Common Mode Input Rejection Ratio	$V_{S-} < V_{CM} < V_{S+}$	90	110		dB
		$T_A = -40^\circ C$ to $+125^\circ C$	80	100		dB
$A_{VOL}$	Open loop voltage gain	$V_{S-} + 0.3V < V_O < V_{S+} - 0.3V$	105	135		dB
		$T_A = -40^\circ C$ to $+125^\circ C$	95			dB
Output characteristics						
$V_{OH}$		$R_L = 10K\Omega$	$(V_{S+}) - 12$	$(V_{S+}) - 3$		mV
		$T_A = -40^\circ C$ to $+125^\circ C$	$(V_{S+}) - 18$			mV
$V_{OL}$		$R_L = 10K\Omega$		$(V_{S-}) + 3$	$(V_{S-}) + 12$	mV
		$T_A = -40^\circ C$ to $+125^\circ C$			$(V_{S-}) + 18$	mV
$I_{SC}$	Short circuit output current	Source current	17	24		mA
		$T_A = -40^\circ C$ to $+125^\circ C$	14			
		Sink current	15	20		mA
		$T_A = -40^\circ C$ to $+125^\circ C$	12			
Power supply characteristics						
PSRR	Power supply rejection ratio	$V_S = 2.3V$ to $5.5V$	90	110		dB
		$T_A = -40^\circ C$ to $+125^\circ C$	80			
$I_Q$	Quiescent current			200	290	$\mu A$
		$T_A = -40^\circ C$ to $+125^\circ C$			380	
Noise characteristics						
$e_n$	Input voltage noise	$f = 0.1Hz$ to $10Hz$		550		nVpp
	Input voltage noise density	$f = 1KHz$		30		nV/ $\sqrt{Hz}$
Dynamic characteristics						
GBW	Gain bandwidth product			2		MHz
SR	Slew rate	$G = \pm 1$		0.8		V/ $\mu s$

## Typical performance characteristics

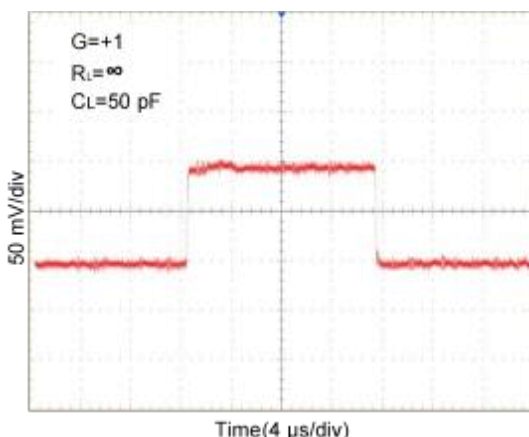
$V_S=+5V$ ,  $T_A=+25^{\circ}C$ ,  $V_{CM}=V_S/2$ ,  $V_O=V_S/2$ ,  $R_L=10k\Omega$  connect to  $V_S/2$ , unless otherwise stated.



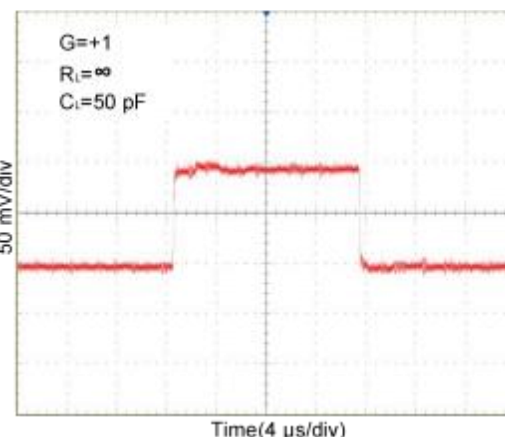
**Large Signal Step Response+5V**



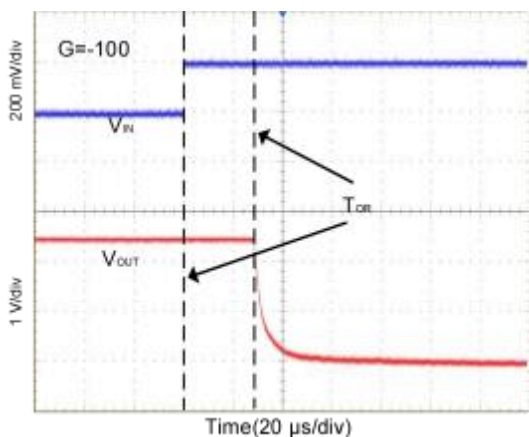
**Large Signal Step Response+2.7V**



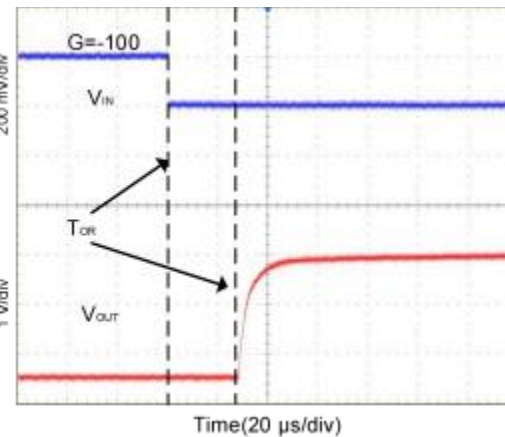
**Small Signal Step Response+5V**



**Small Signal Step Response+2.7V**



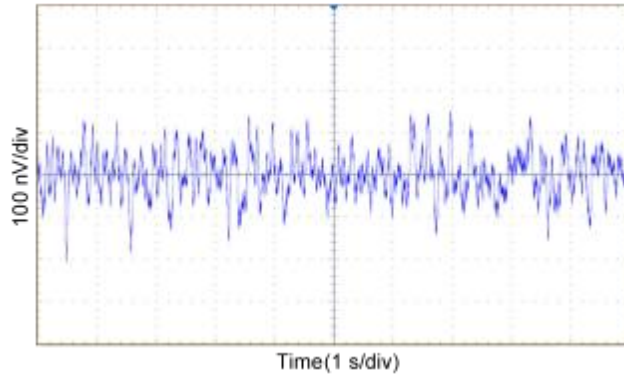
**Positive overload recovery**



**Negative overload recovery**

## Typical performance characteristics

$V_S=+5V$ ,  $T_A=+25^{\circ}C$ ,  $V_{CM}=V_S/2$ ,  $V_O=V_S/2$ ,  $R_L=10k\Omega$  connect to  $V_S/2$  , unless otherwise stated



**0. 1Hz to 10Hz noise**

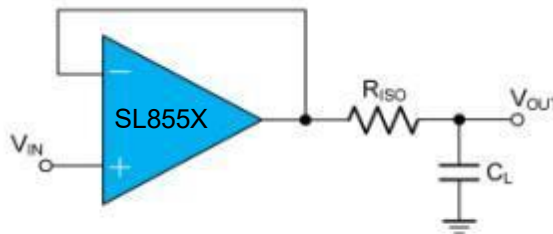
**Application notes**

**1. Working characteristics**

The specified supply voltage of the SL855X series is 2.3V to 5.5V ( $\pm 1.15V$  to  $\pm 2.75V$ ). The specified operating temperature is  $-40^{\circ}C$  to  $+125^{\circ}C$ . Parameters related to supply voltage and temperature can be found in the Classic Characteristics section.

**2. Capacitive loads and their stability**

The unity gain follower (buffer) is the circuit most sensitive to capacitive loads. Directly driving a capacitive load will reduce the phase margin of the operational amplifier, resulting in output ringing or even oscillation. In applications that require a larger capacitive load drive, an isolation resistor RISO needs to be added between the output and the capacitive load, as shown in Figure 1. The isolation resistor RISO and the capacitive load CL will add a zero point, thereby improving stability. The larger the value of RISO, the more stable its output. However, this treatment method will reduce the accuracy of the gain because RISO and the load resistor RL form a voltage divider network.



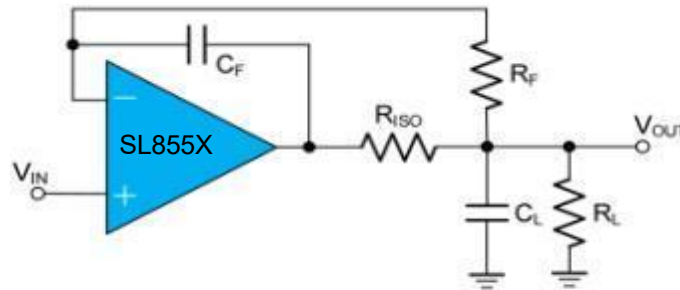
Picture 1. Introduction driving heavy capacitive loads

A better circuit is shown in Figure 2. This circuit has good stability and high DC accuracy. Using an RF resistor to connect the inverting terminal to the output can effectively improve the DC accuracy. CF and RISO are used to compensate for the loss of phase margin. The output signal is fed back to the inverting input through a high-pass element to ensure the phase margin of the overall feedback loop.

For circuits without buffers, there are two other ways to improve the phase margin: 1) increase the gain of the operational amplifier, or 2) prevent a capacitor in parallel with the feedback resistor to compensate for the parasitic capacitance of the inverting input.



## Application notes



Picture 2. Directly drives capacitive loads with high DC accuracy

### 3. Input bias current clock feed through

The SL855X series uses switches to correct the inherent offset and drift of the operational amplifier. However, the internal switch will cause a certain sudden change in the input bias current at the moment of switching. These pulses are very short-lived and are not enough to be amplified by the amplifier, but they can be coupled to the output through the feedback network. The most effective way to prevent this phenomenon is to use a low-pass filter, such as an RC network.

### 4. Layout guide

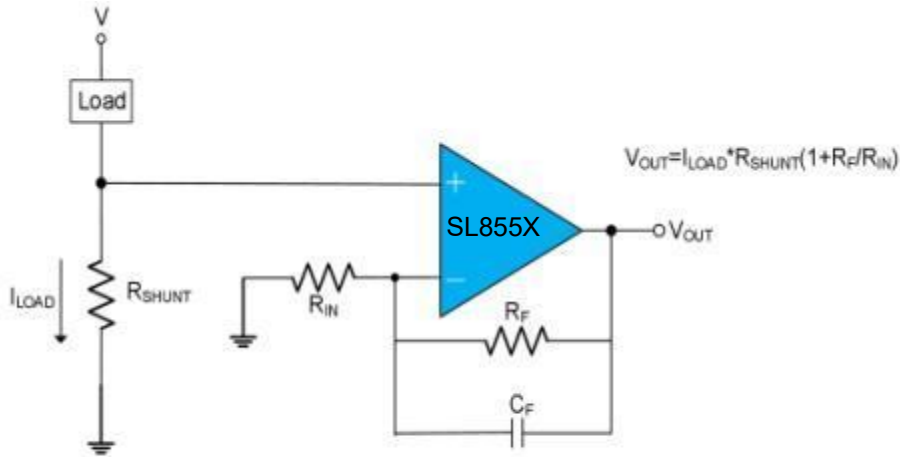
In order to achieve the best performance of the device, the following layout principles should be followed when designing the PCB.

- A. Divide the ground into two parts: digital ground and analog ground, and reasonably plan the path for current to return to the ground to avoid the return of digital signals to analog signals. If a multi-layer PCB is used, set one of the layers as the ground, which not only helps to dissipate heat, but also effectively reduces EMI noise.
- B. In order to minimize the size of parasitic capacitance and Seebeck effect, external devices (such as feedback resistors, etc.) should be as close to the device as possible.
- C. The wire of the input signal should be as short as possible and should be away from the power line or other digital signal lines.
- D. A low ESR, 0.1 $\mu$ F ceramic bypass capacitor should be connected between each power pin and the ground, and as close to the device as possible. In the case of a single power supply, use a capacitor connected between V+ and ground.
- E. Consider adding a low-resistance, driven guard ring around the critical wiring. The guard ring can significantly reduce the leakage current of different potentials nearby.

### 5. Low-side current sensing

**Application notes**

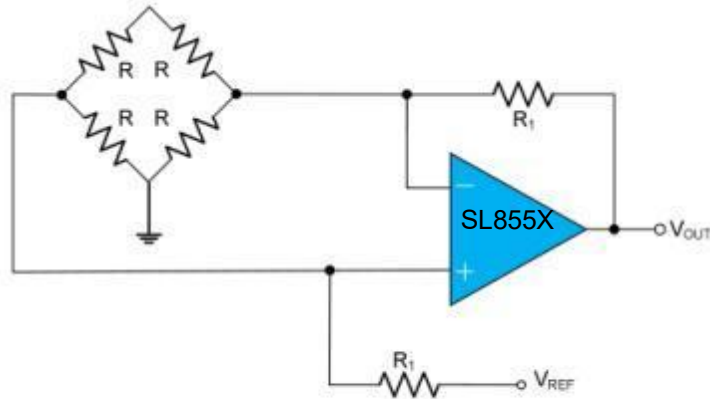
As shown in Figure 3, the operational amplifier forms a low-side current sampling circuit. The load current ( $I_{LOAD}$ ) generates a voltage difference across the resistor  $R_{SHUNT}$  and is amplified by the SL855X. When the power supply voltage remains unchanged, the output voltage range can be changed by changing the resistor  $R_{SHUNT}$  and the closed-loop amplification factor.



Picture 3. Low-side current sensing circuit

**6. Bridge amplifier**

As shown in Figure 4, the SL855X series forms a bridge amplifier.

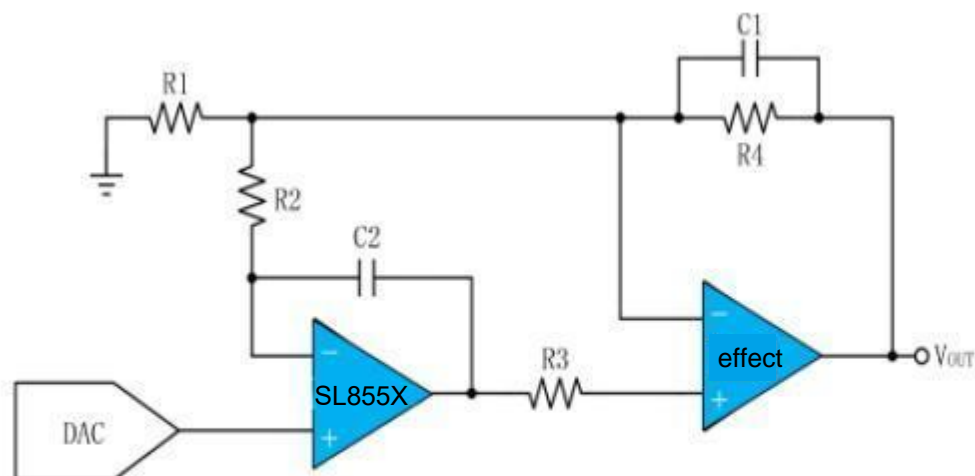


Picture 4. Bridge amplifier

**7. Programmable voltage source**

## Application notes

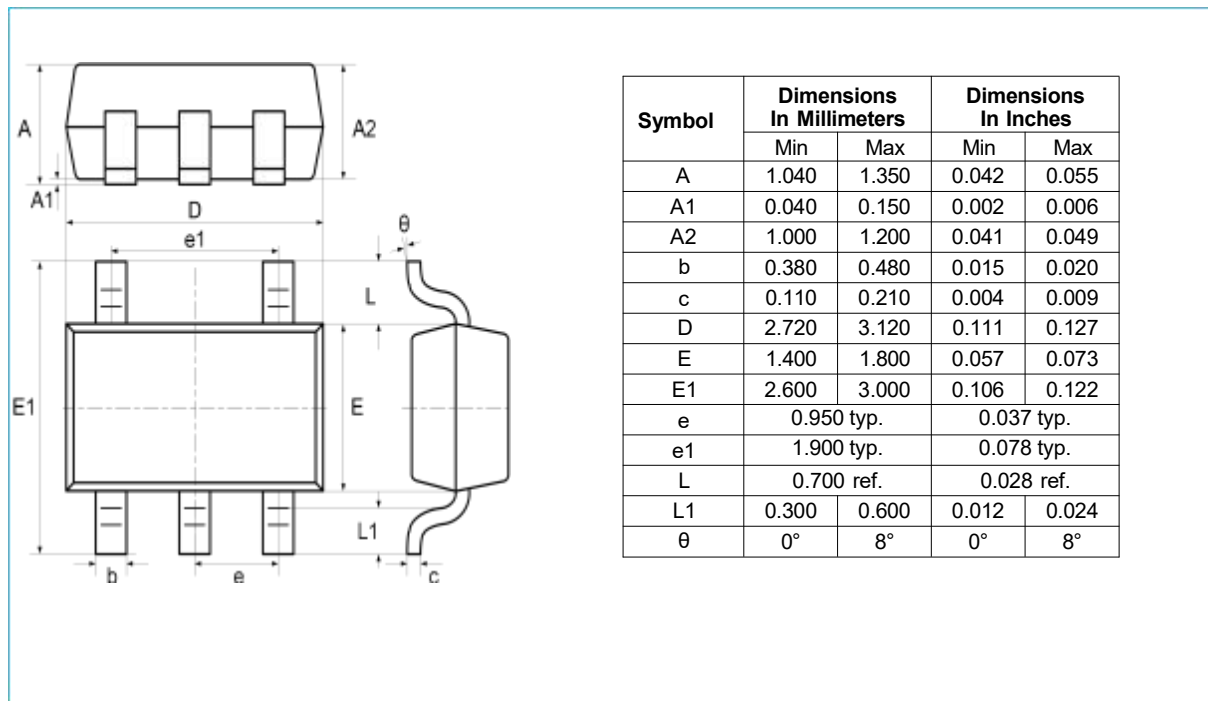
As shown in Figure 5, the SL855X series, DAC, and power amplifier form a high-precision programmable power supply. The amplifier circuit is built using capacitors and resistors to amplify the output voltage of the DAC by a factor of  $1+R4/R1$ . In situations where the input voltage varies over a wide range, the SL855X has the characteristics of high precision and low drift.



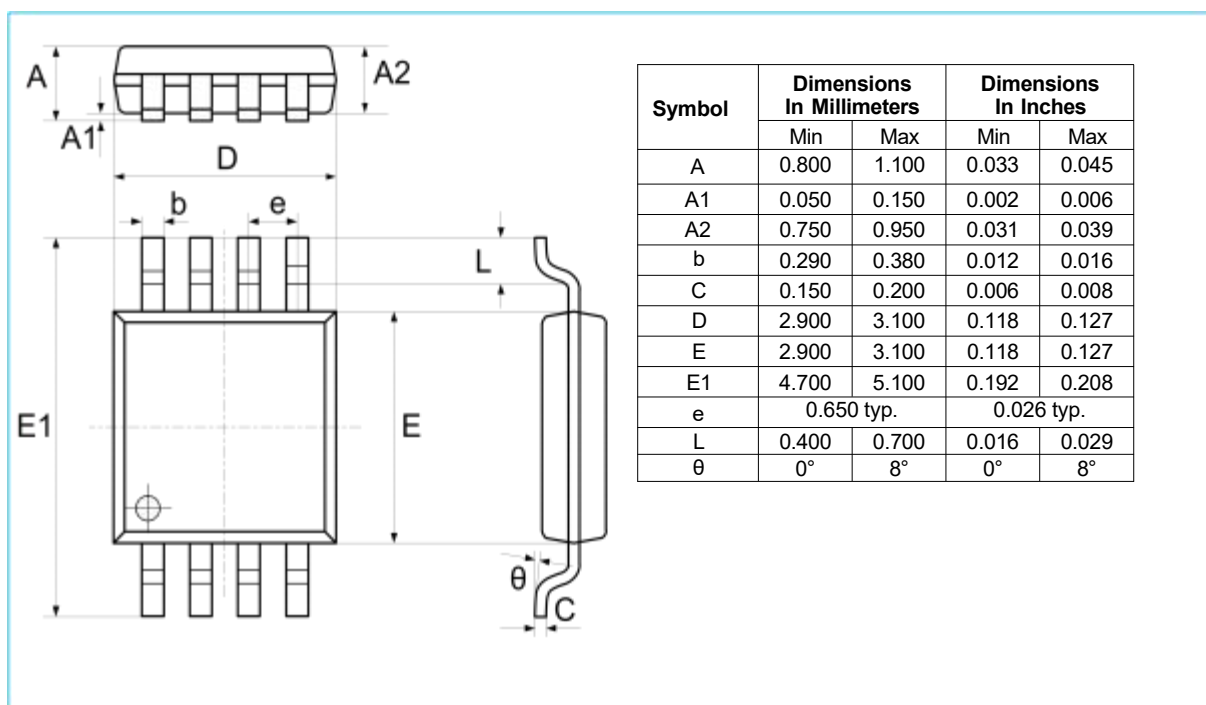
Picture 5. Programmable voltage source

## Package information

### SOT23-5

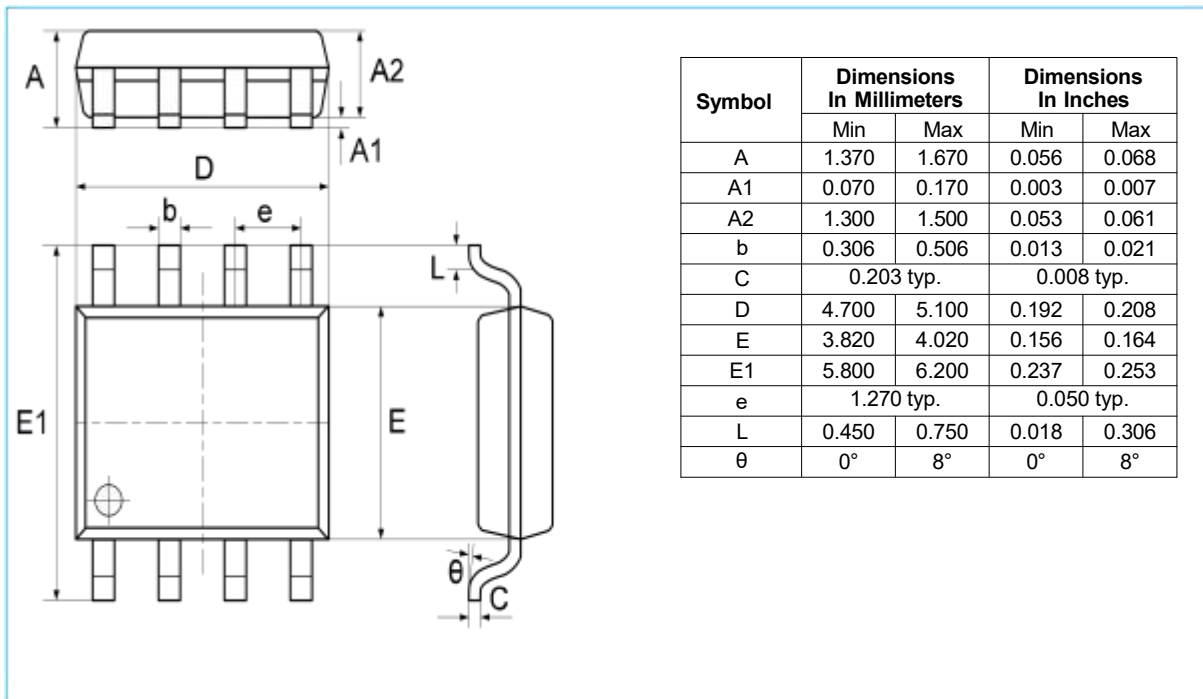


### MSOP-8

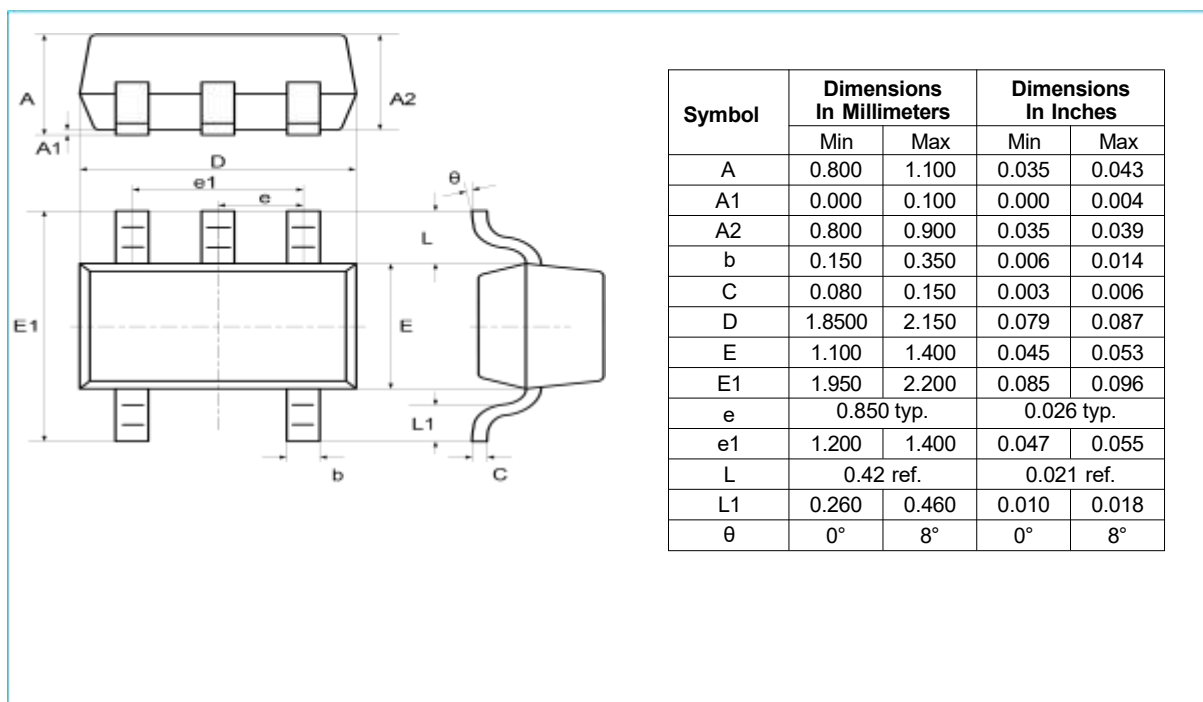


## Package information

### SOP-8



### SC70-5



## Package information

SOP-14

